

### Finding the Right Deep Neural Network Model for Efficient Design of Tunable Nanophotonic Devices

# **Minwoo Jung**<sup>1,2</sup>, Keisuke Kojima<sup>1</sup>, Toshiaki Koike-Akino<sup>1</sup>, Ye Wang<sup>1</sup>, Dayu Zhu<sup>1,3</sup>, and Matthew Brand<sup>1</sup>

<sup>1</sup>Mitsubishi Electric Research Laboratories (MERL), 201 Broadway, Cambridge, MA 02139, USA.

<sup>2</sup>Department of Physics, Cornell University, Ithaca, NY 14853, USA.

<sup>3</sup>School of Electrical and Computer Engineering, Georgia Institute of Technology, Atlanta, GA 30332, USA.

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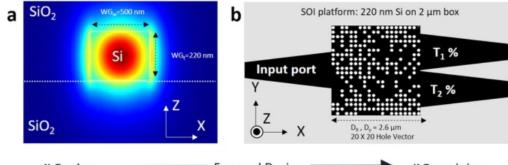
### Neural Networks for Photonic Device Design Application

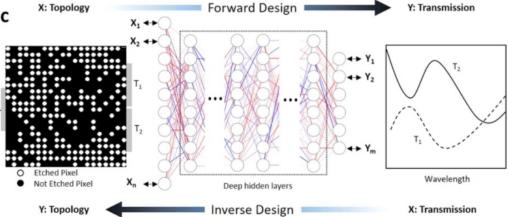
#### **Deep Neural Network Inverse Design of Integrated Photonic Power Splitters**

Mohammad H. Tahersima, Keisuke Kojima 🗁, Toshiaki Koike-Akino, Devesh Jha, Bingnan Wang, Chungwei Lin & Kieran Parsons

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#### **Deep-Learning-Enabled On-Demand Design of Chiral Metamaterials**

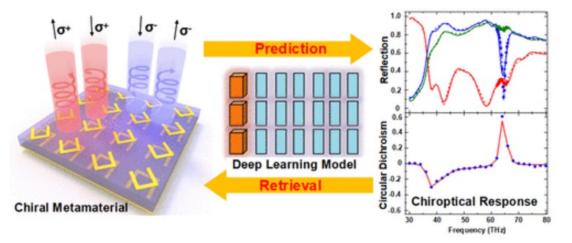
Wei Ma, Feng Cheng, and Yongmin Liu\*

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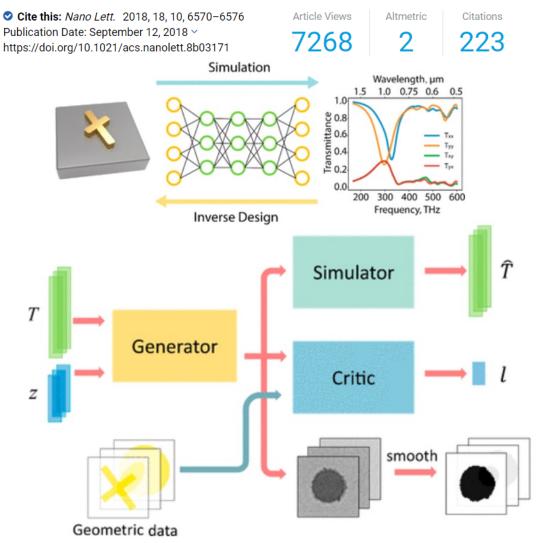
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#### Generative Models : Encoder-Decoder Structure

#### **Generative Model for the Inverse Design of Metasurfaces**

Zhaocheng Liu, Dayu Zhu, Sean P. Rodrigues, Kyu-Tae Lee, and Wenshan Cai\*

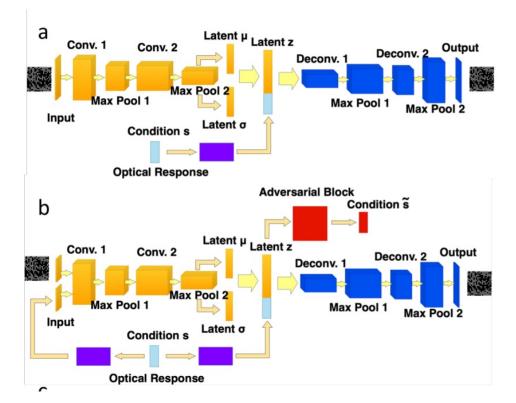




#### Original Paper

### Generative Deep Learning Model for Inverse Design of Integrated Nanophotonic Devices

Yingheng Tang, Keisuke Kojima 🔀, Toshiaki Koike-Akino, Ye Wang, Pengxiang Wu, Youye Xie, Mohammad H. Tahersima, Devesh K. Jha, Kieran Parsons, Minghao Qi,

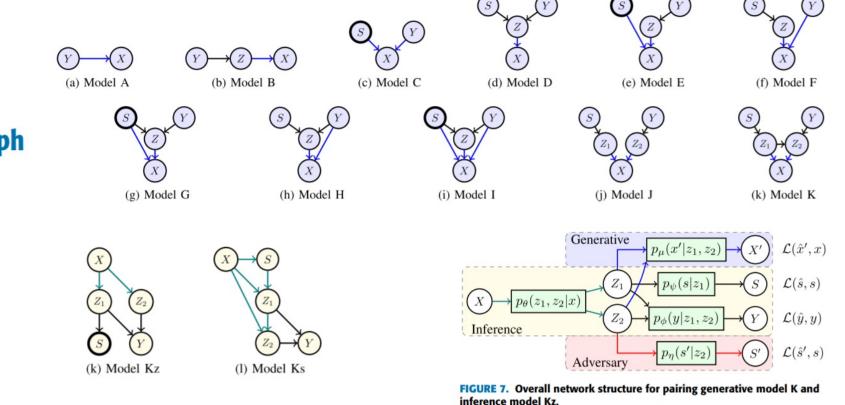


# Bayesian Graph Exploration for finding optimal ANN architecture

#### IEEE Access

#### AutoBayes: Automated Bayesian Graph Exploration for Nuisance-Robust Inference

ANDAC DEMIR<sup>®1</sup>, (Student Member, IEEE), TOSHIAKI KOIKE-AKINO<sup>®2</sup>, (Senior Member, IEEE), YE WANG<sup>®2</sup>, (Senior Member, IEEE), AND DENIZ ERDOGMUS<sup>®1</sup>, (Senior Member, IEEE)



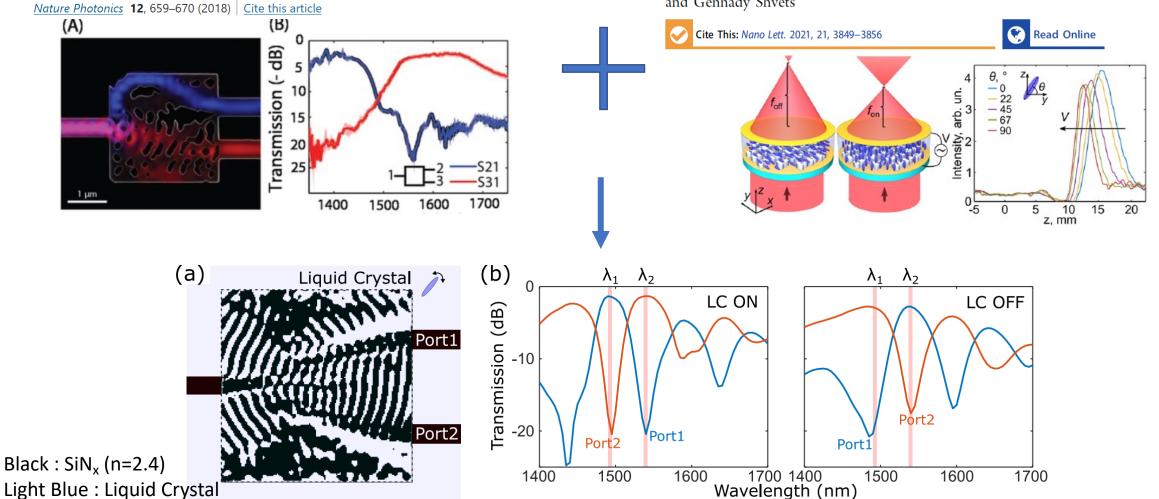
Depending on the specifics of the system, we need to choose the right ANN architecture.

For generative models, we need to come up with the most plausible Bayesian-inference model.

### Considered system : tunable SNOI wavelength splitter

#### Inverse design in nanophotonics

Sean Molesky, Zin Lin, Alexander Y. Piggott, Weiliang Jin, Jelena Vucković & Alejandro W. Rodriguez

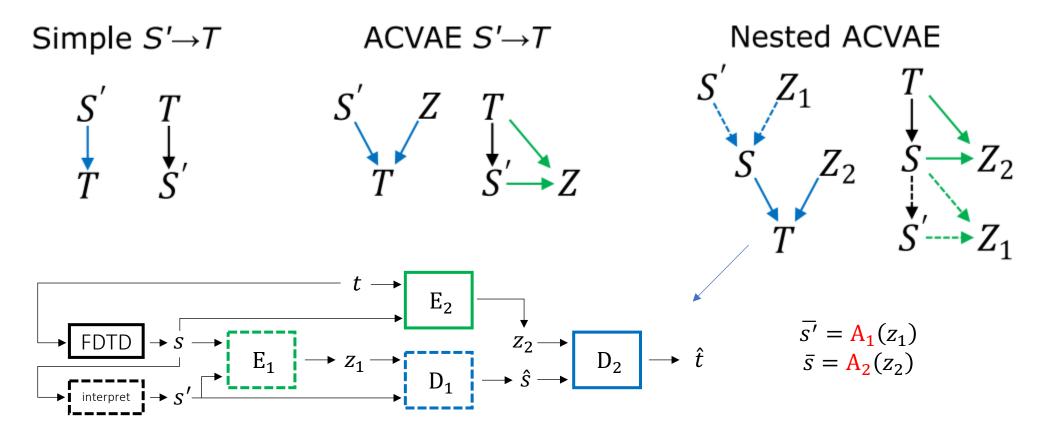


#### Electrically Actuated Varifocal Lens Based on Liquid-Crystal-Embedded Dielectric Metasurfaces

5

Melissa Bosch, Maxim R. Shcherbakov,\* Kanghee Won, Hong-Seok Lee, Young Kim, and Gennady Shvets

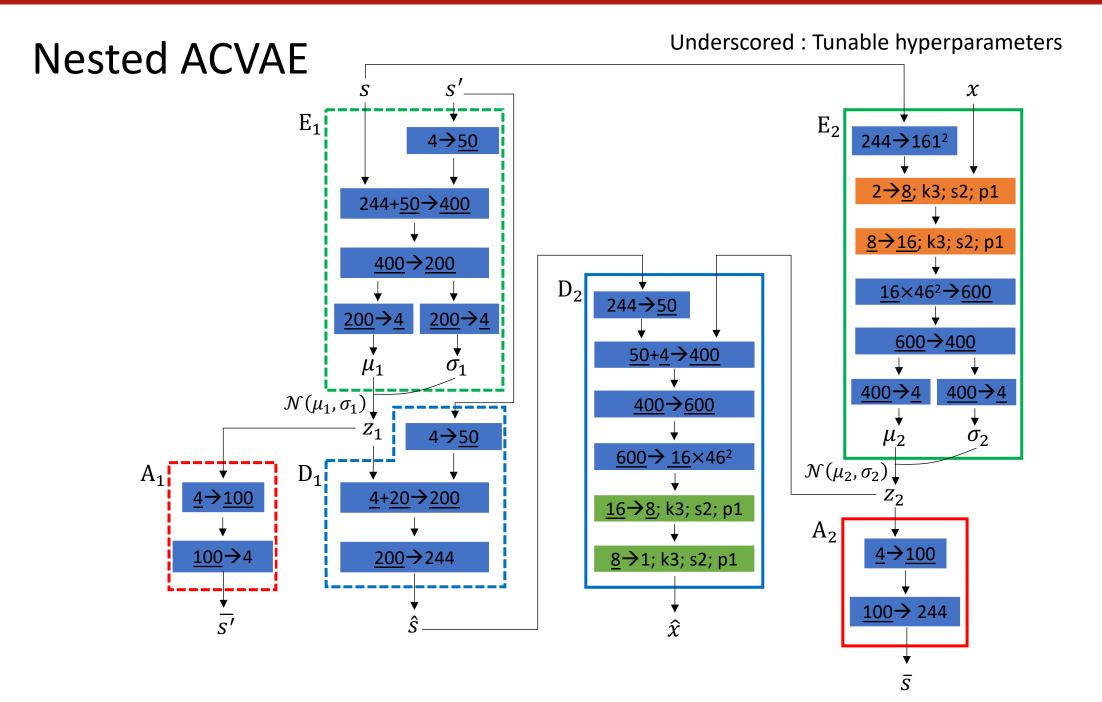
**Considered Bayesian Graphs** 



- t : device topology; 161×161
- *s* : Full spectrum; 4(LC on, off; port 1,2)×61(1.4 $\mu$ m: 5nm: 1.7 $\mu$ m) or 244
- *s*': User-friendly spectrum information;  $[\overline{\lambda}, \Delta\lambda, \overline{ER}]$
- $z_1$ : Latent variable out of the encoder1
- $z_2$ : Latent variable out of the encoder2

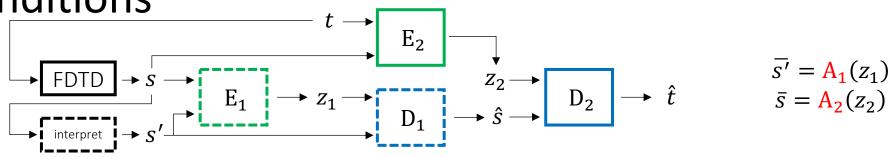
 $\hat{a}$ : Variable a generated out of a decoder

 $\overline{a}$ : Variable a generated out of an adversarial block



#### 

#### Loss conditions



- Basic training losses :  $c_1 MSE(x, \hat{x}) + c_2 MSE(s, \hat{s}) + c_3 KLD(z_1) + c_4 KLD(z_2)$
- Adversarial losses :  $-c_5 MSE(s', \overline{s'}) c_6 MSE(\hat{s}, \overline{s})$
- Cycle-consistency losses :  $c_7 \text{MSE}(z_1, \text{E}_1(\hat{s}; s')) + c_8 \text{MSE}(z_2, \text{E}_2(\hat{t}; s))$
- s'-meaning-enforcing loss :  $c_9MSE(s', I(\hat{s}))$  (*I* refers to the "interpret" dashed box)

+ higher-order-cycle losses : e.g.  $MSE(t, \hat{t})$ , where  $\hat{t} = D_2(z_2 = E_2(\hat{x}, s); \hat{s})$ 

MSE losses can be replaced to any similar losses if necessary. Empirically, I found that  $c_2 MSE(\sqrt{s}, \sqrt{\hat{s}}) + c_9 MSE(\log s', \log I(\hat{s}))$  works better than  $c_2 MSE(s, \hat{s}) + c_9 MSE(s', I(\hat{s}))$ .

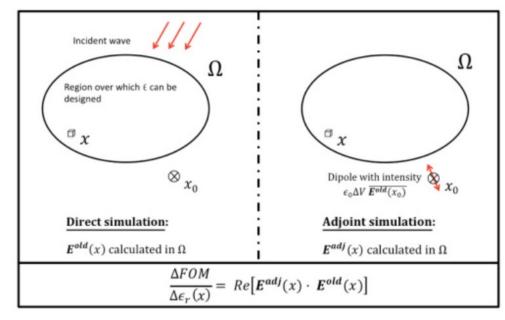
• Ultimate validation loss :  $MSE(s', I(F(\hat{t})))$  (F refers to the "FDTD" solid box)

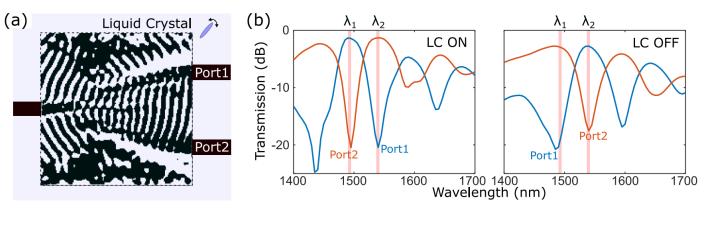
#### Training dataset preparation

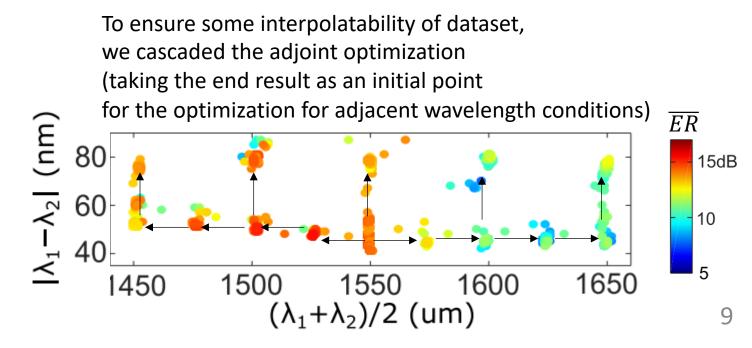
# Adjoint shape optimization applied to electromagnetic design

Christopher M. Lalau-Keraly,<sup>1,\*</sup> Samarth Bhargava,<sup>1</sup> Owen D. Miller,<sup>2</sup> and Eli Yablonovitch<sup>1</sup>

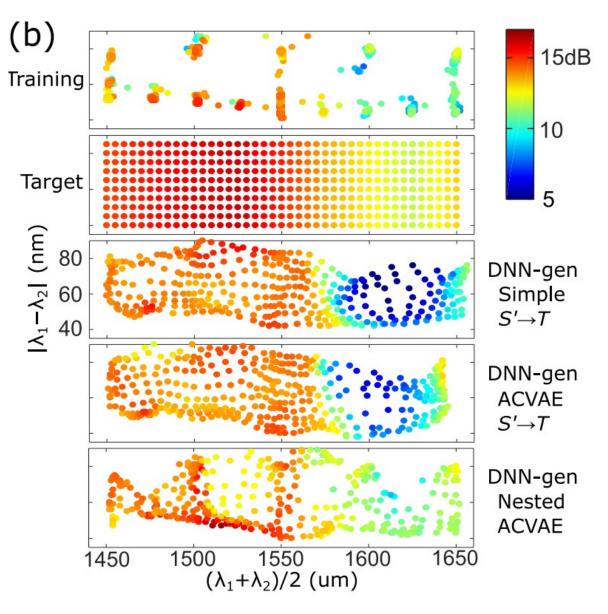
 <sup>1</sup>Department of Electrical Engineering and Computer Sciences, University of California at Berkeley, Berkeley, California 94720, USA
 <sup>2</sup>Department of Mathematics, Massachusetts Institute of Technology, Cambridge, Massachusetts 02139, USA \*chrisker@eecs.berkeley.edu

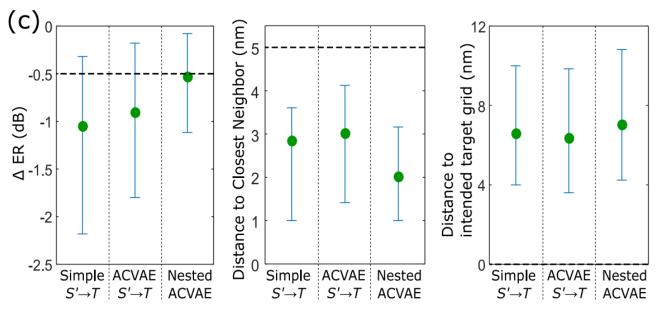






#### Network Validation result



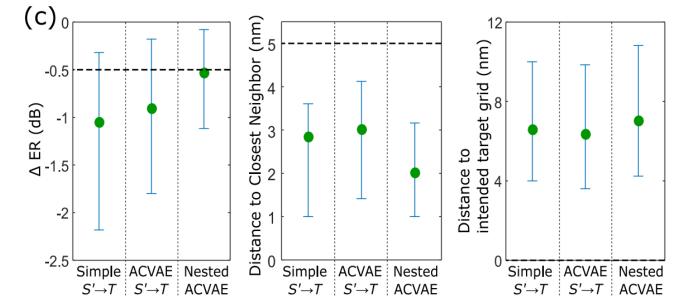


In terms of the extinction ratio of the device, Nested (S'  $S \rightarrow T$ ) ACAVE performs the best

## But, the uniform coverage of optimal device conditions gets a little worse by adding the full spectrum information (S)

\*Time to run a 3D-adjoint optimization : 10~50 hrs
\*Time to train the network : 1~2hrs
\*Time to generate a device topology
from a trained network, and validate in FDTD : 2~3 mins 10

### Discussion



More validations would be needed to draw conclusive remarks, but

- We observed that the inclusion of full spectrum information (S) helps in terms of better generated ER values
- But, it seems that the user-friendly intuitive specs (S';  $[\bar{\lambda}, \Delta\lambda, \overline{ER}]$ ) works better for uniform interpolation

#### **Conclusion and Outlook**

- We demonstrated Auto-Bayes-based network-architecture exploration for optimal design of deep-neural network for complex photonic system (liquid-crystal-tunable wavelength splitter).

- Different architectures show different advantages
- Especially, in a narrow-band wavelength-specific performing photonic devices, the usage of the full spectrum (outside of the wavelength windows that actually matter for the device spec) comes with both plus and minus
- To fully utilize the generative nature of our networks, latent-space optimization (with what values of  $Z_1$  and  $Z_2$  will the generated device performance be optimized?) is an interesting direction to look forward to.

